

Revision of Lecture Two

- **Power lost** (Budget):

Propagation path loss

Slow (large-scale) fading

Fast (small-scale) fading

- Two killer factors in mobile medium:

- **Doppler spread**: time-varying nature of channel causes frequency dispersion

Physical dimension/quantity – Doppler frequency f_D

- **Multipath**: which causes time dispersion

Physical dimension/quantity – excess delay τ

- We will have indepth look into these two phenomena

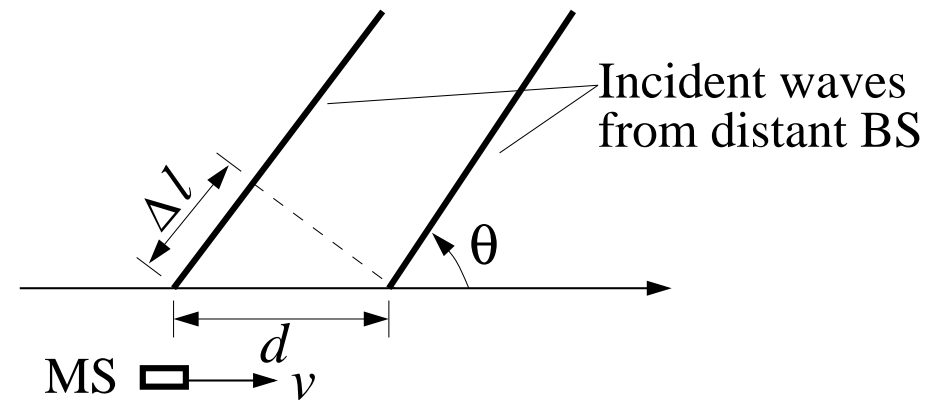


Doppler Frequency

- Consider MS moving at speed v :

Moving “changes” frequency →
Doppler shift

- Difference in path lengths from BS to MS is $\Delta l = d \cos \theta = v \Delta t \cos \theta$



- Let λ be wavelength, then phase change in received signal due to difference in path lengths is:

$$\Delta\phi = \frac{2\pi\Delta l}{\lambda} = \frac{2\pi v\Delta t \cos\theta}{\lambda}$$

- Doppler frequency** is defined as rate of phase change due to moving:

$$f_D = \frac{1}{2\pi} \frac{\Delta\phi}{\Delta t} = \frac{v}{\lambda} \cos\theta = f_m \cos\theta$$

where f_m is the maximum Doppler frequency (unit: Hz)

Doppler Spectrum

- The arrival angle θ can be viewed as uniformly distributed

$$\text{PDF}(\theta) = \frac{1}{2\pi}, \quad 0 \leq \theta \leq 2\pi$$

The Doppler frequency $f_D = f_m \cos \theta$ is then cosine distributed

- Received power in $d\theta$ around θ is proportional to $\frac{|d\theta|}{2\pi}$ (Use of absolute operation is due to fact that power is always nonnegative). Using

$$\frac{d \cos^{-1} x}{dx} = -\frac{1}{\sqrt{1-x^2}}, \quad -1 \leq x \leq 1$$

- Doppler power spectrum density:** (absolute operation because power ≥ 0)

$$S(f_D) \propto \frac{1}{2\pi} \left| \frac{d\theta}{df_D} \right| = \frac{1}{2\pi} \left| \frac{d(\cos^{-1}(f_D/f_m))}{df_D} \right| \quad \text{or} \quad S(f_D) = \frac{C}{\sqrt{1 - (f_D/f_m)^2}}$$

- Implications: **frequency dispersion**

- Single frequency f_c broadened to a spectrum ($f_c - f_m, f_c + f_m$)
- Signal with bandwidth $2B$ centre at f_c broadened to a bandwidth approximately $2B + 2f_m$



Doppler Spread

- **Doppler spread** B_D is defined as the “bandwidth” of Doppler spectrum. It is a measure of spectral broadening caused by the time varying nature of the channel
- **Coherence time** $T_C \propto \frac{1}{B_D}$ is used to characterise the time varying nature of the frequency dispersion of the channel in time domain
- **Fading** effects due to Doppler spread: determined by mobile speed and signal bandwidth. Let baseband signal bandwidth be B_S and symbol period T_S , then
 - “Slow fading” channel: $T_S \ll T_C$ or $B_S \gg B_D$, signal bandwidth is much greater than Doppler spread, and effects of Doppler spread are negligible
 - “Fast fading” channel: $T_S > T_C$ or $B_S < B_D$, channel changes rapidly in during one symbol period T_S

Do not confuse with slow (large-scale) and fast (small-scale) fadings in propagation pathloss model

Here slow and fast fading are used to describe relationship between **time rate of change** in the **channel** and the transmitted **signal**

Normalised Doppler Frequency

- Velocity of mobile and signal bandwidth determine whether a signal undergoes fast or slow fading
- Fading rate describes the relationship between rate of change in channel and rate of change in signal
 - **Rate of change in channel** is specified by velocity of mobile v and carrier frequency f_c , as characterised in the (maximum) Doppler frequency

$$f_m = \frac{v}{\lambda} = \frac{v \cdot f_c}{c}, \quad \lambda \text{ being wavelength, } c \text{ being speed of light}$$

- As signal bandwidth is much smaller than f_c , Doppler spread is approximately f_m
- **Rate of change in signal** is specified by symbol rate or symbol period T_s
- Often **normalised Doppler frequency** is used to specify fading rate

$$\bar{f}_m = f_m \cdot T_s$$

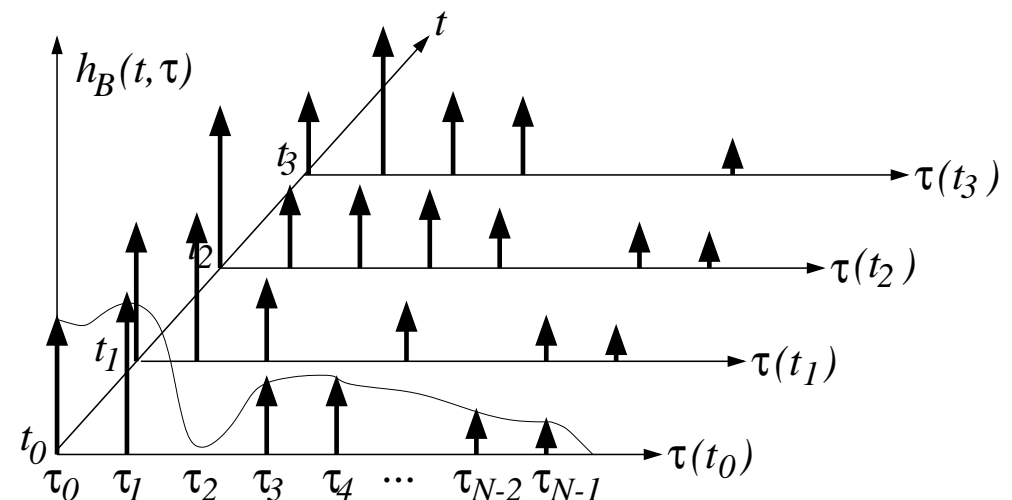
- $\bar{f}_m = 10^{-6}$ is considered very slow fading, $\bar{f}_m = 10^{-4}$ quite fast
- Example: Carrier frequency of 1 GHz \rightarrow wavelength $\lambda = c/f_c = 3 \cdot 10^8 / 10^9 = 0.3$ m
User velocity of 10 m/s (36 km/h) and $\lambda = 0.3$ m \rightarrow Doppler frequency $f_m = v/\lambda \approx 33$ Hz
At symbol rate of 3.3 Msymbols/s, the normalised Doppler frequency becomes $\bar{f}_m = f_m \cdot T_s = 33 / (3.3 \cdot 10^6) = 10^{-5}$

Impulse Response of Multipath Channels

- **Multipath** causes **time dispersion**, as described by bandpass CIR $h(t, \tau)$
 - As channel can be time-varying, **time** t is needed, and τ is **multipath delay**
 - Generally, $h(t, \tau)$ is a function of two inputs t and τ
- Let equivalent baseband **complex-envelope** channel impulse response be $h_B(t, \tau)$

$$h_B(t, \tau) = \sum_{i=0}^{N-1} a_i(t, \tau) \exp(-j\theta_i(t, \tau)) \delta(\tau - \tau_i(t))$$

- Useful to discrete τ into delay bins, each bin represents a multipath component
- $a_i(t, \tau)$, $\theta_i(t, \tau)$ and $\tau_i(t)$ are **amplitude**, **phase shift** and **excess delay** of i th multipath component, respectively



Channel Impulse Response (continue)

- Interpretation of $h_B(t, \tau)$: there are N multipaths, i.e. there are N copies of transmitted signal arriving at the receiver
- At time t , each copy arrives at the receiver with a different **amplitude** $a_i(t, \tau)$, goes through a different **phase shift** $\theta_i(t, \tau)$ and has a different **excess delay** $\tau_i(t)$
- Excess delay τ is function of t , amplitude is function of t and τ , phase shift is function of t and τ , and they are stochastic processes
- A special case is the time invariant channel, where

$$h_B(\tau) = \sum_{i=0}^{N-1} a_i \exp(-j\theta_i) \delta(\tau - \tau_i)$$

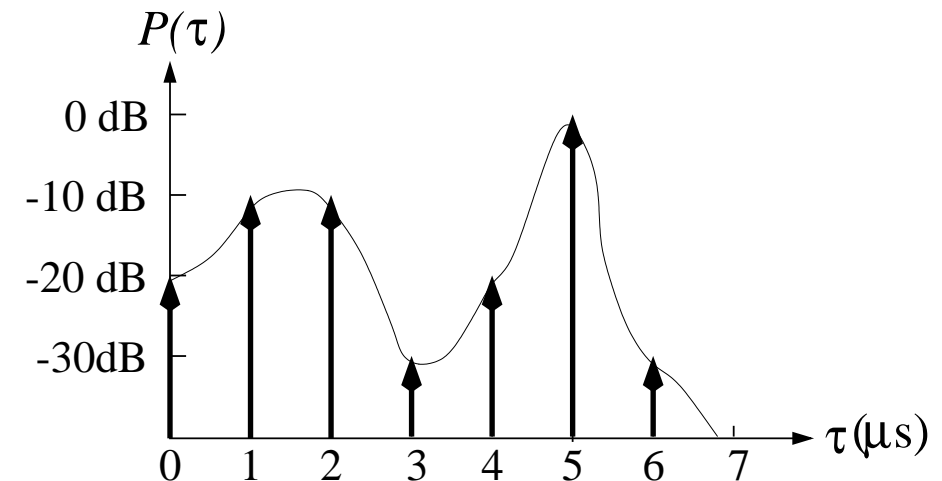
θ_i : uniformly distributed, a_i : Rayleigh distributed, τ_i : Poisson distributed

- **Multipath** causes **time dispersion** and results in **intersymbol interference**



Power Delay Profile

- **Power delay profile** $P(\tau)$: the channel power spectral density as a function of delay, i.e. how “channel power” is distributed along **dimension excess delay** τ
- Consider a local area around a spatial position, averaging $|h_B(t, \tau)|^2$ over time gives rise to $P(\tau)$
- Specifically, $P(\tau)$ is Fourier transform of autocorrelation function of $h_B(t, \tau)$
- Again, it is useful to discrete τ into bins
- **Mean excess delay** is defined as the first moment of power delay profile



$$\bar{\tau} = \frac{\sum_i P(\tau_i) \tau_i}{\sum_i P(\tau_i)}$$

Power delay profile or power spectral density has “properties” of probability density function, so one can talk about moments of the underlying “stochastic process”

Power Delay Profile (continue)

- **Root mean square (RMS) delay spread** is defined as the square root of the second central moment of power delay profile:

$$\sigma_{\tau} = \sqrt{\overline{\tau^2} - (\overline{\tau})^2}$$

where the second moment is given by

$$\overline{\tau^2} = \frac{\sum_i P(\tau_i) \tau_i^2}{\sum_i P(\tau_i)}$$

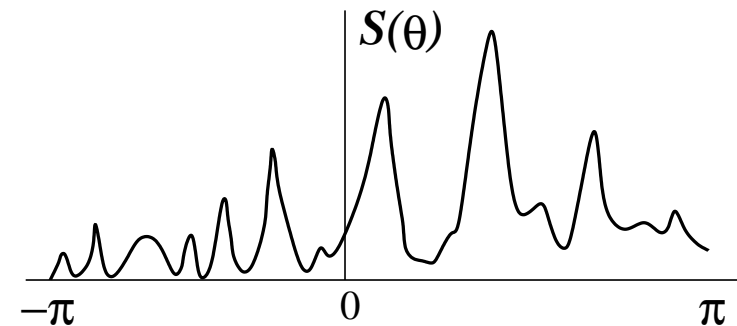
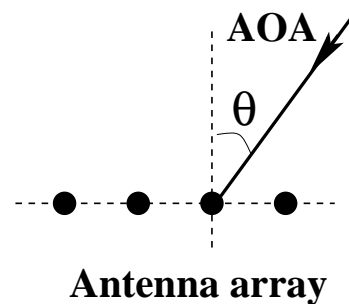
- **Coherence bandwidth** is a measure of the range of frequencies over which the channel is “flat” (i.e. passing spectral components with approximately **equal gain and linear phase**)
- 50% coherence bandwidth is defined as:

$$B_C \approx \frac{1}{5\sigma_{\tau}}$$



Angle Power Spectrum

- **Angle power spectrum** defines average power as a function of θ (angle-of-arrival for receive antenna and angle-of-departure for transmit antenna)

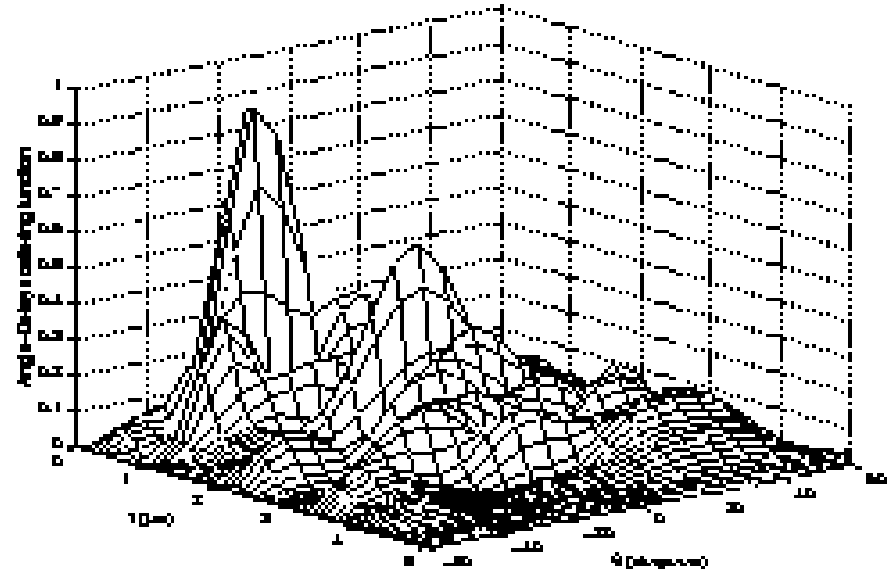
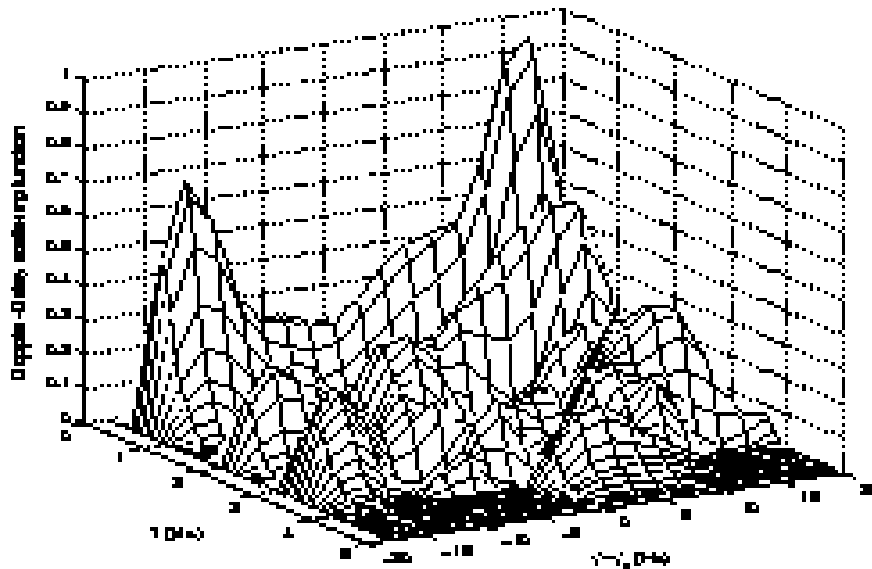


- Similar to delay power profile, one can define Mean angle $\bar{\theta}$ and RMS angle spread σ_{θ}
- **Angle spread** causes **space selective fading** \rightarrow signal amplitude depends on spatial location of antenna/signals
- **Coherence distance** D_C is spatial separation for which autocorrelation coefficient of spatial fading drops to 0.7

$$D_c \propto \frac{1}{\sigma_{\theta}}$$

Scattering Functions

- Doppler-delay and **angle-delay scattering functions** $S(f_D, \tau)$ and $S(\theta, \tau)$



- Complete channel statistics are captured in a triple scattering function: **Doppler-angle-delay scattering function** $S(f_D, \theta, \tau)$
- Doppler, delay and angle power spectra $S(f_D)$, $P(\tau)$ and $S(\theta)$ are **marginal spectra** related to scattering function $S(f_D, \theta, \tau)$

Summary

- Mobile channels are hostile due to:
 - Doppler spread which causes frequency dispersion
 - Multipath which causes time dispersion
- Doppler spectrum (note speed broadens signal spectrum):
 - Doppler PSD, Doppler spread, coherence time
 - What are **slow** and **fast** fading channels
 - Normalised Doppler frequency
- Multipath: channel impulse response and power delay profile
 - mean excess delay, RMS delay spread, coherence bandwidth
- Doppler-angle-delay scattering function, and marginal spectra
 - angle power spectrum, RMS angle spread, coherence distance

