## **Revision of Lecture Eleven**

- Previous lecture we have concentrated on carrier recovery for QAM, and modified early-late clock recovery for multilevel signalling as well as star 16QAM scheme
- Thus we have completed Modem, under ideal AWGN or flat fading channel condition



If channel is dispersive, equalisation is required  $\Rightarrow$  we will return to this issue as well as issue of multiple access later

 $\bullet$  We now turn to CODEC part  $\Rightarrow$  we will concentrate on channel coding and decoding, but not source coding and decoding

For practical source coder and decoder, refer to 2nd half unit of Digital Transmission

## **Channel Coding Introduction**

- Mobile channels are very hostile environments, and yet real systems work satisfactorily. One of the contributors to this success is channel coding
- Channel coding is used to detect and often correct symbols that are received in error
- Error detection can be used by receiver to generate ARQ to transmitter for a re-transmission of the frame in error, as in computer networks (stop & wait, go-back-*n*, selective repeat protocols)
- When re-transmission is not an option: **forward error correction** coding, which introduces extra information (redundancy) into transmitted data for receiver to detect and correct errors

		Convolutional codes						
Others								
	non-cyclic		Polynomial (cy					
		Golay	Bose-Chaudhuri					
			Reed-Solomon					

Some examples:

- Binary BCH and Convolutional codes widely used in various practical communications systems
- Reed-Solomon codes used in music CD
- Golay codes used in Mars explorer



## **Block Code Introduction**

• There are systematic and non-systematic codes. For block codes, systematic ones are more powerful

Rate R = k/n block code: k information bits plus r = n - k check bits forms a **codeword**. All valid codewords form a **codebook** 

• (n,k) systematic block code



Systematic: k information bits must be explicitly transmitted (more strict definition also requires they are transmitted together as a block)

• Systematic linear block code: first k bits of a codeword are message bits, and last n-k check bits are linear combinations of the k message bits



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### Linear Block Code: Encoding

- Let  $\mathbf{c}$  be *n*-bit codeword and  $\mathbf{d}$  be *k*-bit message, written in row-vector form
- An (n,k) linear block code is defined by its  $k \times n$  generating matrix G

$$G = [I_k \mid P]$$

with  $k\times (n-k)$  matrix P specifying the given (n,k) linear block code, and  $I_k$  being identity matrix of order k

• Encoding process can then be written as

$$\mathbf{c} = \mathbf{d}G$$

• All elements in P are binary valued, and binary (modulo-2) arithmetic operations are carried out

A binary sequence of n bits should have  $2^n$  patterns, denoting as  $\bar{\mathbf{c}}_i$ ,  $1 \le i \le 2^n$ , but c only contains  $2^k$  codewords, i.e. it can only take some of  $\{\bar{\mathbf{c}}_i\}$ , called **legal** sequences  $\rightarrow$  only these legal sequences can be transmitted

If receiver encounters an illegal sequence  $\bar{\mathbf{c}}_i$  (not a codeword), what it says?

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#### Example

• (6,3) linear block code with generating matrix and codebook

$$G = \begin{bmatrix} 1 & 0 & 0 & | & 0 & 1 & 1 \\ 0 & 1 & 0 & | & 1 & 0 & 1 \\ 0 & 0 & 1 & | & 1 & 1 & 0 \end{bmatrix}$$

massages	codewords
000	000 000
001	001 110
010	010 101
011	011 011
100	100 011
101	101 101
110	110 110
111	111 000

• For example, for message **d**=110, parity check bits are

$$c_4 = 1 \cdot 0 + 1 \cdot 1 + 0 \cdot 1 = 0 + 1 + 0 = 1$$
  

$$c_5 = 1 \cdot 1 + 1 \cdot 0 + 0 \cdot 1 = 1 + 0 + 0 = 1$$
  

$$c_6 = 1 \cdot 1 + 1 \cdot 1 + 0 \cdot 0 = 1 + 1 + 0 = 0$$

Note the binary modulo-2 arithmetic operations involved

- $2^6 = 64$  but only  $2^3 = 8$  legal codewords e.g. 111111 is not a legal codeword
- If receiver encounters 111111 it must be due to error, as 111111 will never be sent





#### Linear Block Code: Decoding

• Each  $k \times n$  generating matrix  $G = [I_k \mid P]$  is associated with a  $(n - k) \times n$  parity check matrix

$$H = [P^T \mid I_{n-k}]$$

Basic property of codeword: c is a codeword in the (n, k) block code generated by G, if and only if  $cH^T = 0$ 

 $\bullet\,$  Received row vector  ${\bf r}$  can be written as

 $\mathbf{r} = \mathbf{c} + \mathbf{e}$ 

All the elements are binary valued, e.g. if the transmitted  $c_i = 1$  and is received in error:  $r_i = 0$ , then  $e_i = 1$ 

• (n-k) (row vector) error syndrome

$$\mathbf{s} = \mathbf{r}H^T = (\mathbf{c} + \mathbf{e})H^T = \mathbf{c}H^T + \mathbf{e}H^T = \mathbf{e}H^T$$

 ${\bf s}$  is related to the error vector  ${\bf e},$  and can be used to detect and correct errors



#### **Error Detection and Correction Capabilities**

- Weight of a codeword c is the number of nonzero elements in c
- Hamming distance between two codewords  $c_1$  and  $c_2$  is the number of elements in which they differ
- Minimum distance of a codebook,  $d_{\min}$ , is the smallest Hamming distance between any pair of codewords in the codebook
- The minimum distance  $d_{\min}$  of a linear block code is equal to the minimum weight of any nonzero codeword in the code
- Code with  $d_{\min}$  can detect up to  $d_{\min} 1$  errors and correct up to  $(d_{\min} 1)/2$ errors in each codeword

Here we are considering binary codes, where Hamming distance is defined

For error correction capability, we refer to hard-input hard-output decoding, i.e. decoder input is in hard bits and it outputs hard bits, later we will see soft-input decoding has better capability





#### **Cyclic Codes**

- Cyclic or polynomial generated codes are subset of linear block codes with some nice properties
- Definition of cyclic: if  $(c_0, c_1, \dots, c_{n-2}, c_{n-1})$  is a codeword then  $(c_{n-1}, c_0, \dots, c_{n-3}, c_{n-2})$  is also a codeword in the same code
- A k-bit message  $\mathbf{d} = (d_0, d_1, \cdots, d_{k-1})$  can be described by a message polynomial d(x):

$$d(x) = d_0 + d_1 x^1 + \dots + d_{k-1} x^{k-1}$$

• The code is defined by its generating polynomial

$$g(x) = g_0 + g_1 x^1 + \dots + g_r x^r$$
 with  $g_0 = 1$  and  $g_r = 1$ 

• The *n*-bit codeword  $\mathbf{c} = (c_0, c_1, \cdots, c_{n-1})$  for  $\mathbf{d}$  is described by a polynomial

$$c(x) = \operatorname{Rem}\left(\frac{x^r \cdot d(x)}{g(x)}\right) + x^r \cdot d(x)$$

where the remainder of  $x^r \cdot d(x)/g(x)$ ,  $\operatorname{Rem}(x^r \cdot d(x)/g(x))$ , is a polynomial up to order  $x^{r-1}$  (i.e. r check bits), called **parity check** polynomial for d(x)

• All calculations use modulo-2 arithmetic

# Cyclic Codes (continue)

• Example of (7,4) cyclic code with  $g(x) = 1 + x^2 + x^3$ : for message d = 0101,  $d(x) = x^1 + x^3$ ,  $x^3 \cdot d(x) = x^4 + x^6$ ,  $\operatorname{Rem}(x^3 \cdot d(x)/g(x)) = 1$ ,  $c(x) = 1 + x^4 + x^6$ , and thus

	check	message					
$\mathbf{c} =$	$1 \ 0 \ 0$	$0\ 1\ 0\ 1$					

• In decoding, the received r(x) = c(x) + e(x) with nonzero terms in e(x) indicating errors, and the **syndrome** polynomial is calculated:

$$\operatorname{Rem}\left(\frac{c(x)+e(x)}{g(x)}\right) = \operatorname{Rem}\left(\frac{e(x)}{g(x)}\right) = s(x)$$

- If it is a zero syndrome: no error or undetectable errors (e(x) contains factor g(x)); if a nonzero syndrome: errors detected and it is used for error correction
- Encoding and syndrome calculation can easily be implemented using shift register feedback circuits



## Cyclic Code Encoder

• (n,k) cyclic code encoder: an (n-k) stage shift register with a feedback circuit



- The circuit operates under a clock and an encoding cycle consists of  $\boldsymbol{n}$  shifts
  - Shift register always starts at zero state, i.e. all  $r_i = 0$ , and ends at zero state
  - During the first k shifts, S1 is closed  $\rightarrow$  shift d(x) into the shift register; and S2 is down  $\rightarrow$  copy d(x) directly to c(x)
  - After the k-th shift, the contents of the (n-k) stage shift register are the n-k parity check bits for d(x)
  - During the remaining n-k shifts, S1 is open and S2 is up  $\rightarrow$  clear the shift register contents out to c(x)





(7,4) cyclic code with  $g(x) = 1 + x + x^3$ Given message  $d(x) = 1 + x^2 + x^3$ :



input		shift	register			codeword									
				index	$r_0$	$r_1$	$r_2$	$c_0$	$c_1$	$c_2$	$c_3$	$c_4$	$c_5$	$c_6$	
1	0	1	1	0	0	0	0	_			_	_	_	_	
	1	0	1	1	1	1	0	-	-	-	-	-	-	1	
		1	0	2	1	0	1	_	_	_	-	_	1	1	
			1	3	1	0	0	_	_	_	-	0	1	1	
			_	4	1	0	0	_	_	_	1	0	1	1	
			_	5	0	1	0	_	_	0	1	0	1	1	
			-	6	0	0	1	_	0	0	1	0	1	1	
			_	7	0	0	0	1	0	0	1	0	1	1	



# **Cyclic Code Syndrome Calculation**

• (n,k) cyclic code syndrome calculation circuit:



• The register is initialised to the zero state

S1 is closed and S2 is opened  $\rightarrow$  the received r(x) is shifted into register

After this, contents of register are s(x)

S1 is opened and S2 is closed  $\to s(x)$  is shifted out and the register is cleared, ready for the next cycle



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## **Other FEC Codes**

• BCH is a subset of cyclic codes with the largest  $d_{\min}$  for given (n,k) and a BCH code is denoted by  $(n,k,d_{\min})$ . This is a class of powerful and widely used FEC codes

Non-binary (i.e. can take values not just  $0 \mbox{ and } 1$ ) version is called Reed-Solomon code and is used e.g. in music CD

- Golay codes: e.g. Mars explorer uses Golay code
- Convolutional codes:

In block codes, a *n*-bit codeword at a time unit t, c(t), depends only on the k-bit data, d(t), at the time t

For convolutional codes,  ${\bf c}(t)$  also depends on N~(N>0) previous blocks of data  ${\bf d}(t-i),~1\leq i\leq N$ 

CC(n,k,N): rate  $R=k/n{\rm ,}$  constraint length N (or memory  $N+1{\rm )}{\rm ,}$  usually  $n{\rm ,}k$  and N are small



## Summary

- Channel coding introduction: FEC coding and classification
- Systematic block codes ⊃ linear block codes ⊃ cyclic (polynomial generated) codes
   ⊃ binary BCH codes

Error detection and correction capabilities

- Systematic linear block codes: generating matrix and encoding; parity check matrix and syndrome
- Cyclic codes: how every things can be described by polynomials, encoder and syndrome calculation (shift register feedback circuits)

BCH: subset of cyclic codes with the largest  $d_{\min}$  for given (n,k)

• Convolutional codes: differences with linear block codes